

following elegant result:

$$H(M, R) = H(M, M + 1 - R).$$

For instance, in Table 1 of the above letter<sup>1</sup> we have  $H(6, 2) = H(6, 5) = 1.9729$ .

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## A Note on the Sampling of Zero Crossings of Two-Dimensional Signals

AVIDEH ZAKHOR AND DAVID IZRAELEVITZ

*Curtis et al. applied a theorem due to Bezout to show that almost all continuous, periodic, band-limited two-dimensional signals can be reconstructed from at most  $4(N_1 + N_2)^2$  zero-crossing samples where  $N_1$  and  $N_2$  is the number of Fourier coefficients in the signal. In this letter we prove a new version of Bezout's theorem and apply it to the above problem to provide a more lenient sampling requirement of at most  $8N_1N_2$  zero-crossing samples.*

#### INTRODUCTION

A considerable amount of research in the field of communication theory has been devoted to the problem of reconstruction of signals from their zero crossings [1]–[3]. Recently, Curtis *et al.* [1] applied Bezout's theorem to the problem of reconstructing a real, band-limited, continuous-time periodic two-dimensional signal  $h(x, y)$  from a finite number of its zero crossings. Specifically if  $F(n_1, n_2)$ , the Fourier series coefficients of the signal, have a rectangular region of support given by

$$\begin{aligned} -N_1 &\leq n_1 \leq N_1 \\ -N_2 &\leq n_2 \leq N_2 \end{aligned}$$

then almost all  $h(x, y)$  can be uniquely determined from at most  $4(N_1 + N_2)^2$  samples of its zero crossings. We will show that because of the geometry of the problem at most  $8N_1N_2$  zero crossings are required and in the process we develop a tighter version of an important theorem in algebraic geometry, Bezout's theorem.

#### A TIGHTER VERSION OF BEZOUT'S THEOREM

Bezout's theorem is concerned with determining the number of common zeros of two bivariate polynomials and can be stated in the following manner:

*Theorem 1 [5]: If two bivariate polynomials of degree  $r$  and  $s$  given by*

$$\begin{aligned} f(x, y) &= \sum_{i=0}^r \sum_{j=0}^{r-i} a(i, j) x^i y^j \\ g(x, y) &= \sum_{i=0}^s \sum_{j=0}^{s-i} b(i, j) x^i y^j \end{aligned}$$

*have no common factor of degree greater than zero, then there exists at most  $rs$  common zeros.*

A band-limited, continuous-time periodic signal  $h(x, y)$  with

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periods  $T_1$  and  $T_2$  in  $x$  and  $y$  can be represented in terms of a polynomial  $h'(w, z)$  in the variables  $w = e^{j(2\pi x/T_1)}$  and  $z = e^{j(2\pi y/T_2)}$

$$h(x, y) = w^{N_1} z^{N_2} h'(w, z)$$

$$h'(w, z) = \sum_{n_1=0}^{2N_1} \sum_{n_2=0}^{2N_2} F(n_1 - N_1, n_2 - N_2) w^{n_1} z^{n_2}. \quad (1)$$

Therefore using Bezout's theorem, if  $h'(w, z)$  is irreducible, then  $h'(w, z)$  (or equivalently  $h(x, y)$ ) can be uniquely determined from  $4(N_1 + N_2)^2$  zeros of  $h'(w, z)$  (or  $h(x, y)$ ). This is because the degree of  $h'(w, z)$  in  $w$  and  $z$  is  $2(N_1 + N_2)$  and Bezout's theorem deals with total degree rather than degree in each variable. In other words, it is concerned with polynomials whose coefficients have triangular support as shown in Fig. 1. On the other hand, one is usually interested in images with square or rectangular support in the Fourier domain where many of the coefficients corresponding to a triangular support are zero as shown in Fig. 2.

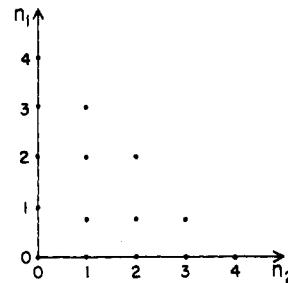


Fig. 1. Triangular support of a polynomial of degree 4.

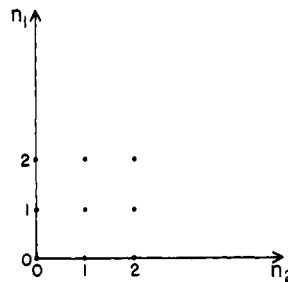


Fig. 2. Square support of a polynomial of maximum degree 2 in  $x$  and maximum degree 2 in  $y$ .

For the case when the polynomials under consideration have rectangular support, we are able to lower the bound on the number of common finite zeros from the bound set by Bezout's theorem. Specifically, if  $f$  and  $g$  are given by

$$f(x, y) = \sum_{i=0}^{N_x} \sum_{j=0}^{N_y} a(i, j) x^i y^j \quad (2)$$

$$g(x, y) = \sum_{i=0}^{M_x} \sum_{j=0}^{M_y} b(i, j) x^i y^j \quad (3)$$

the upper bound on the number of common finite zeros set by Bezout's theorem is  $(N_x + N_y)(M_x + M_y)$ . Our objective is to establish a tighter upper bound on the number of common finite zeros of  $f$  and  $g$ .

Before proceeding, we need to review several results concerning the resultant of polynomials in one or two variables. The resultant  $R_{pq}$  of two polynomials  $p$  and  $q$  in a single variable  $x$

$$p(x) = a_0 + a_1 x + a_2 x^2 + \dots + a_N x^N$$



## CONCLUSION

We found an upper bound  $N_x M_y + M_x N_y$  for the number of common finite zeros of two relatively prime polynomials given by (4) and (5). This bound is much tighter than  $(N_x + N_y)(M_x + M_y)$  which is obtained via Bezout's theorem. For example, if the coefficients of a polynomial have region of support of  $N \times N$ , then it can be uniquely determined using  $2N^2$  samples of its finite zero crossings as opposed to  $4N^2$ .

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# On the Realization of Two's Complement Overflow Limit Cycle Free State-Space Digital Filters

VIMAL SINGH

A frequency-domain criterion for the overflow stability of a class of state-space digital filters is presented. The result is derived using a Lyapunov technique and the associated MKY machinery. Some consequences of the present derivation are discussed.

## I. INTRODUCTION

Barnes and Fam [1] presented a "minimum norm" criterion for the absence of overflow oscillations in a class of state-space digital filters. Mills, Mullis, and Roberts [2] presented a generalized version of the criterion of [1]. Specifically, if there exists a positive diagonal matrix  $D$  such that  $D - A^T D A$  is positive definite, then overflow oscillations are avoided. Kawamata and Higuchi [3] have carried out a Lyapunov analysis to derive a criterion which is analogous to the criterion of [2]. More recently, Bolton [4] has discussed realization, using Lyapunov method, of a two's complement overflow limit cycle free digital filter structure (of order two).

The purpose of this letter is to point out, employing a Lyapunov technique and the associated MKY machinery, a frequency-domain criterion for the overflow stability of the above-referred state-space digital filters. Some consequences of the present derivation are discussed in Section IV.

## II. FREQUENCY-DOMAIN CRITERION

The state-variable filter under consideration is of the form [1]-[4]

$$x(r+1) = f(y(r)) \quad (1a)$$

$$y(r) = Ax(r) \quad (1b)$$

where  $x(r)$  is an  $n$ -vector state,  $A$  is the  $n \times n$  coefficient matrix, and  $f(y(r))$  is an  $n$ -vector nonlinear function, namely,

$$f(y(r)) = \begin{bmatrix} f_1(y_1(r)) \\ f_2(y_2(r)) \\ \vdots \\ f_n(y_n(r)) \end{bmatrix} \quad (2)$$

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i.e., the  $n$ -vector  $y(r)$  stands for the  $n$ -vector  $Ax(r)$ . It is assumed that

$$\det(zI - A) \neq 0, \quad \text{for all } |z| \geq 1 \quad (3)$$

and that

$$\begin{aligned} (A, I) &: \text{completely controllable;} \\ (A, A) &: \text{completely observable} \end{aligned} \quad (4)$$

where  $I$  is the  $n \times n$  identity matrix.

Assume a nonlinearity of the form

$$f_i(0) = 0, \quad |f_i(y_i(r))| \leq |y_i(r)|, \quad i = 1, 2, \dots, n. \quad (5)$$

The characteristic given in (5) includes, among others, an important arithmetic, namely, two's complement [1]-[4].

A criterion for the overflow stability of the above-described filter is given in the following theorem.

**Theorem:** (A sufficient condition for the absence of overflow oscillations.) For the zero solution of the circuit described by (1)-(5) to be asymptotically stable in the large, it is sufficient that there exists a positive diagonal matrix  $C$  such that the following is satisfied:

$$C + CA(zI - A)^{-1} + [CA(zI - A)^{-1}]^* \geq 0, \quad \text{for all } |z| = 1 \quad (6)$$

where "\*" denotes the conjugate transpose and " $\geq$ " signifies that the matrix is positive semidefinite.

## III. DERIVATION OF THE STABILITY CRITERION

Consider a quadratic Lyapunov function

$$v(x(r)) = x^T(r) P x(r) \quad (7)$$

where  $P = P^T$  is positive definite and  $T$  denotes the transpose. Application of (7) and (1a) results in

$$\begin{aligned} \Delta v(x(r)) &= v(x(r+1)) - v(x(r)) \\ &= f^T(y(r)) P f(y(r)) - x^T(r) P x(r). \end{aligned} \quad (8)$$

Adding to and subtracting from (8) the quantity

$$[f(y(r)) + y(r)]^T C [f(y(r)) - y(r)] \quad (9)$$

where  $C$  is a positive diagonal matrix, yields

$$\begin{aligned} \Delta v(x(r)) &= -x^T(r) [P - A^T C A] x(r) - f^T(y(r)) [C - P] f(y(r)) \\ &\quad + [f(y(r)) + y(r)]^T C [f(y(r)) - y(r)] \end{aligned} \quad (10)$$

where (1b) has been utilized. With the substitution

$$f(y(r)) = y(r) - \tilde{f}(y(r)) \quad (11)$$

and employing (1b), (10) can be expressed as

$$\begin{aligned} \Delta v(x(r)) &= -x^T(r) [P - A^T P A] x(r) - x^T(r) [A^T P - A^T C] \tilde{f}(y(r)) \\ &\quad - \tilde{f}^T(y(r)) [P A - C A] x(r) \\ &\quad - \tilde{f}^T(y(r)) [C - P] \tilde{f}(y(r)) \\ &\quad + [f(y(r)) + y(r)]^T C [f(y(r)) - y(r)]. \end{aligned} \quad (12)$$

Set

$$P - A^T P A = Q Q^T \quad (13a)$$

$$P A - C A = R Q^T \quad (13b)$$

$$C - P = R R^T \quad (13c)$$

where  $Q$  and  $R$  are real  $n \times n$  matrices. Then (12) takes the form